DYNAMIC ANALYSIS OF A NEW CONNECTION FOR DUAL VOLTAGE OPERATION OF SINGLE PHASE CAPACITOR RUN MOTOR

H. H. Hanafy
Electrical power and machines Dept., Faculty of Engineering, Cairo University, Giza-Egypt.
Add.: Electrical power and machines dept., Faculty of Engineering, Cairo University, Giza-Egypt, Zip 12613.
E-mail: Hanafy_Hassan@hotmail.com

Abstract: This paper presents the dynamic analysis of a new connection of Single-Phase Capacitor Run Motor (SPCRM) for dual voltage operation. A program simulating the dynamic behavior of the new connection in the stationary d-q reference frame gives sufficient results. The analysis is carried out under the no-load and load conditions. Experimental measurements are carried out for comparison and verification of simulating results. The agreement of experimental and computational results approves the dynamic model of the new connection.

Key words: Single phase induction motor, capacitor run motor, dynamic analysis, dual voltage.

Nomenclature
- $r_{1m}$ main winding resistance;
- $r_{1a}$ auxiliary winding resistance;
- $r_s$ rotor windings resistance referred to the main winding;
- $L_m$ magnetizing inductance referred to the main winding;
- $L_{d0}$ main winding leakage inductance;
- $L_{d0}$ auxiliary winding leakage inductance;
- $L_{r0}$ rotor windings leakage inductance referred to the main winding;
- $i_m$ main current;
- $i_a$ auxiliary current;
- $i_{ar}$ rotor current of phase α;
- $i_{br}$ rotor current of phase β;
- $i_{dr}$ direct component of rotor current;
- $i_{qr}$ quadrature component of rotor current;
- $v_m$ main voltage;
- $v_a$ auxiliary voltage;
- $v_C$ voltage across the capacitor;
- $P$ number of pole-pairs;
- $J$ moment of inertia (kg·m²);
- $B_f$ viscous friction coefficient (N·m/(rad/s))
- $\omega$ electrical angular velocity of the rotor;
- $T_L$ load torque;
- $n$ motor speed;
- $p$ differential operator w.r.t. time.

1. Introduction

Single-phase induction motors are widely used because of their cost, reliability and robustness in operation. Single phase capacitor motors operate either as “capacitor run motor”, where the auxiliary winding is permanently energized, or as “capacitor start motor”, where the auxiliary winding is switched out of circuit, or finally as “capacitor start capacitor run motor”, where the value of the capacitor is large during starting and small during running.

The single phase capacitor run motors (SPCRM) are used for many applications such as heating, ventilating and air conditioning (HVAC) blowers and compressors. The SPCRMs have many advantages such as moderated starting torque and good running performance (high output power, high power factor, high efficiency and low supply current)[1].

The dual voltage operation of the SPCRMs is common in Europe, Australia, most regions of Africa and Asia and few countries in South America use the voltage of (220/240 V, 50 Hz), while most of the North American countries use (110/127 V, 60 Hz).

The author in [2] presented the theory of operation and the steady state analysis of a new connection of the SPCRMs (as shown in Fig. 1) for dual voltage operation as well as for supplying the SPCRMs with the line-to-line voltage instead of the line-to-neutral voltage of the same grid in case of missing the neutral point, and showed that the suggested connection has a satisfactory performance and overcomes the disadvantages of the conventional dual voltage connections given in [3-4].

Through this paper the transient analysis of the new connection will be carried out. The transient analysis of the electric motors is very essential because it helps in determining the insulation strength, suitability of winding, shaft strength, value of capacitor, and devising the protection strategy.
Dynamic simulation of electrical machines is performed by several methods, and it is mostly based on the d-q model [5]. Tichenor et al. [6] presented a multiple reference frame based model of a generically configured permanent split capacitor induction machine. The model was valid for both steady state and transient analysis. In [7] the authors applied the dynamic phasor methodology to modeling of a single-phase induction machine with a position dependent load. The developed dynamic model of the machine predicted its transients well, and could be used in high performance control algorithms. Sadowski et al. [8] carried out the dynamic operation of a single phase induction motor using a general method coupling field and electrical circuit equations. Simulation of a single-phase induction machine using 2-D finite elements method was presented in [9]. Also there are many available software packages for the analysis of the electric machines such as EMTP and PSPICE. The authors in [10] investigated the transient phenomena of capacitor induction machines, where the machine equations were evaluated using a power system package Electromagnetic Transients Program (EMTP) Universal Machine (UM) modeling. Jawad et al. [11] investigated the dynamic Analysis of a single-phase induction motor using PSPICE and compared the results with that obtained by EMTP, showing the advantages of using PSPICE over EMTP.

This paper presents the dynamic analysis of the new connection of the SPCRM for dual voltage operation under no load and load conditions. The simulation results obtained were proved experimentally in the laboratory by loading the SPCRM using a dynamometer.

2. Mathematical Model

The mathematical model of the SPCRM will be constructed in the stationary d-q frame. Fig. 2 shows the configuration of the SPCRM, where the squirrel cage rotor may be represented as a pair of symmetrical and orthogonal, short-circuited coils (α, β). Since the axes of the stator windings are orthogonal, the d-q axes of the model may be aligned with axes of the windings, as illustrated in Fig. 2. The stator variables remain unchanged because there is no transformation on these variables. The transformation used to transform the rotor variables to the stationary d-q frame is given in (1).

\[
\begin{bmatrix}
  i_{dr} \\
  i_{dq}
\end{bmatrix} = \begin{bmatrix}
  \cos \theta & \sin \theta \\
  -\sin \theta & \cos \theta
\end{bmatrix} \begin{bmatrix}
  i_{pL} \\
  i_{cL}
\end{bmatrix}
\]

(1)

The dynamic model of the new connection of the SPCRM in the stationary d-q frame can be described by set of differential equations, written in the matrix form as follows:

\[
\begin{bmatrix}
  v_m \\
  v_c
\end{bmatrix} = \begin{bmatrix}
  r_{oL} + pL_{oL} & 0 & pL_o & 0 \\
  0 & r_{rL} & 0 & pL_o \\
  0 & pL_o & \frac{\alpha}{pL_o} & r_p + pL_o & \frac{\alpha}{pL_o} \\
  0 & -\frac{\alpha}{pL_o} & pL_o & -\frac{\alpha}{pL_o} & r_p + pL_o
\end{bmatrix} \begin{bmatrix}
  i_{dr} \\
  i_{dq}
\end{bmatrix}
\]

(2)

\[
v_i = v_m + v_a \quad , \quad v_u = v_c
\]

(3)

\[
i_a = i_c + i_a
\]

(4)

\[
i_c = C p v_a
\]

(5)

The mechanical equations which represent the dynamics of the mechanical system can be written as:

\[
T_{em} = pL_o (i_{dr} i_{dq} - i_a i_c)
\]

(6)

\[
p\omega = \frac{p}{J} (T_{em} - B_1 \omega - T_L)
\]

(7)

\[
p\dot{\theta} = \omega
\]

(8)

Where:

\[
L_{um} = L_{dm} + L_o
\]

\[
L_{si} = \alpha^2 L_{dm} + L_o
\]

\[
L_{ei} = L_o + L_{dm}
\]

\[
i_a = a_i \alpha
\]

α = the effective turns ratio of the auxiliary winding to the main winding.

\[T_{em}\] = the electromechanical torque of the motor.

θ = angular displacement of the rotor w.r.t. the stator.
Simulation Results

The dynamic simulation of the new connection is carried out using a capacitor run single-phase induction motor with the following specifications: 175 W, 220 V, 50 Hz, 1440 rpm, 1.5 A, and a 10 µF capacitor. Table 1 gives the experimentally determined parameters of the motor.

The operation of the new connection is studied by a supply voltage of 380 V-50 Hz to the motor configuration shown in Fig. 1.

Table 1 Parameters of the SPCRM

<table>
<thead>
<tr>
<th>Parameter</th>
<th>Value</th>
</tr>
</thead>
<tbody>
<tr>
<td>$r_{lm}$</td>
<td>26 Ω</td>
</tr>
<tr>
<td>$r_{la}$</td>
<td>24 Ω</td>
</tr>
<tr>
<td>$L_{lm}$</td>
<td>0.0871 H</td>
</tr>
<tr>
<td>$L_{la}$</td>
<td>0.067 H</td>
</tr>
<tr>
<td>$r_r$</td>
<td>14.32 Ω</td>
</tr>
<tr>
<td>$L_{lr}$</td>
<td>0.087 H</td>
</tr>
<tr>
<td>$L_o$</td>
<td>0.9022 H</td>
</tr>
<tr>
<td>$J$</td>
<td>0.0035 Kg.m$^2$</td>
</tr>
<tr>
<td>$B_f$</td>
<td>0.003 N.m/(rad/s)</td>
</tr>
<tr>
<td>$a$</td>
<td>1.0235</td>
</tr>
</tbody>
</table>

3.1. No-Load Conditions

The simulation results in this section are obtained for free acceleration of the motor with zero load torque. Only the moment of inertia of the motor “J” and the viscous friction coefficient “$B_f$” were taken into account.

Fig. 3 shows the speed time curve, where the acceleration period ends at approximately one sec. and the motor reaches its no load speed of 1490 rpm.

Figs. 4-5 show the main winding and the auxiliary winding currents with time; it is clear that the currents decrease to their steady state values as the motor attains the no load speed. Figs. 6-7 represent the main winding and the auxiliary winding voltages versus time; it is clear that the voltages remain nearly constant at low speeds until the speed has reached the steady state value then they increase to their steady state value as the motor attains the no load speed.

Fig. 8 illustrates the rotor phase current against time; it is obvious that the rotor current has high values and frequency during the acceleration period and low value and very low frequency at steady state due to the operating slip.

Fig. 9 shows the variation of the electromagnetic torque of the motor; it can be observed very high oscillations of torque at twice supply frequency which influence on pulsations of the speed. These oscillations are caused by elliptical rotating field due to the phase difference between the main and auxiliary windings currents and also the unequal amplitudes of these currents. The elliptical shape of the rotating magnetic field can be observed in the trajectory of the stator currents shown in Fig. 10.
Fig. 5 The auxiliary current versus time at no load

Fig. 6 The main voltage versus time at no load

Fig. 7 The auxiliary voltage versus time at no load

Fig. 8 The rotor phase current versus time at no load

Fig. 9 The electromagnetic torque versus time at no load

Fig. 10 Trajectory of the stator currents at no load
3.2. Load Conditions

The simulation results in this section are obtained at load conditions by applying a load torque of 1.0 N·m (approximately 85% of the full load torque) after 2.5 sec.

Fig. 11 shows the speed time curve, it can be noticed, that the speed decreases from the no load value of 1490 rpm to 1430 rpm at load conditions.

Figs. 12-13 show the main winding and the auxiliary winding currents with time; it can be observed that the main current increases with loading, while the auxiliary current slightly decreases.

Figs. 14-15 represent the main winding and the auxiliary winding voltages versus time; it is clear that the main voltage remains nearly constant with loading, while the auxiliary voltage decreases.

Fig. 16 demonstrates the rotor phase current against time; it is obvious that the amplitude and the frequency of the rotor current increase with loading due to the high operating slip. Also it can be noticed, that the rotor current has oscillations, which are caused by the pulsations of the speed due to oscillations of the electromagnetic torque.

Fig. 17 shows the variation of the electromagnetic torque of the motor. The electromagnetic torque is composed of the steady state time averaged torque and the double frequency pulsating torque. The pulsating torque produces noise and vibration during running, while the average torque is responsible for overcoming the friction and driving the load. This double-stator frequency torque oscillations are produced by the interaction of oppositely rotating mmf waves which relatively rotate to each other at twice the synchronous speed.
4. Experimental Results

The new connection of the SPCRM is tested and in the laboratory by mechanically coupling the SPCRM to a dynamometer, where the load torque could be electrically controlled. The speed is recorded using a DC tachometer with a gain of 1V/500 rpm. The currents and voltages waveforms are recorded using dual-channel digital oscilloscope through current and voltage transducers. Table 2 gives the specifications of the transducers.

Table 2 Specifications of the transducers

<table>
<thead>
<tr>
<th>Transducer</th>
<th>Input signal</th>
<th>Output signal</th>
</tr>
</thead>
<tbody>
<tr>
<td>Current</td>
<td>0-5 A</td>
<td>0-10 V</td>
</tr>
<tr>
<td>Voltage</td>
<td>0-600 V</td>
<td>0-10 V</td>
</tr>
</tbody>
</table>

4.1. No Load Results

The experimental results in this section are obtained for free acceleration of the motor with zero load torque. Fig. 18 shows the speed time curve. Figs. 19-20 show the main winding and the auxiliary winding currents with time. Figs. 21-22 give the main winding and the auxiliary winding voltages versus time.
Fig. 20 The experimental auxiliary current versus time at no load (2.5 A/Div. & 0.25 Sec./Div)

Fig. 21 The experimental main voltage versus time at no load (120 V/Div. & 0.25 Sec./Div)

Fig. 22 The experimental auxiliary voltage versus time at no load (120 V/Div. & 0.25 Sec./Div)

4.2. Load Results

The experimental results in this section are obtained at load conditions by applying a load torque of 1.0 N-m; after the motor reaches its steady state no load speed. Figs. 23-24 show the main winding and the auxiliary winding currents with time; it can be observed that the main current increases with loading, while the auxiliary current slightly decreases.

Figs. 25-26 represent the main winding and the auxiliary winding voltages versus time; it is clear that the main voltage remains nearly constant with loading, while the auxiliary voltage decreases.
By comparing the results of sections 3 & 4; one can conclude that, the simulations waveforms are very close to those obtained from the experimental results. The small differences between the simulation and the experimental results can be explained as a result of neglecting the core losses, effect of the temperature and effect of the saturation in the mathematical dynamic model.

5. Conclusions
In this paper, the dynamic analysis of a new connection for operating the SPCR from dual voltage, in the stationary d-q frame, has been presented. Experimental work has been carried out to examine the proposed mathematical dynamic model for the new connection of the SPCR and to ensure the accuracy of computational technique under different operating conditions. Experimental results show correctness and validity of the presented mathematical dynamic model of the SPCR.

The results show that all voltages and currents of the motor windings are within the safe limits during the starting, no load and load conditions.

References